

Finite Sample Analysis of AMP

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Statistical Physics and Machine Learning Back Together,
August 2018

Joint work with Adam Greig, Kuan Hsieh, and Ramji
Venkataramanan (Cambridge)

High-Dimensional Regression

High-dimensional regression: $y = A\beta + w$,

$$\begin{bmatrix} y \end{bmatrix} = \begin{matrix} m \\ \left[\begin{array}{c} \\ \\ \end{array} \right] \end{matrix} \begin{matrix} \overbrace{}^N \\ \left[\begin{array}{c} \\ \\ \end{array} \right] \end{matrix} \begin{bmatrix} \beta \end{bmatrix} + \begin{bmatrix} w \end{bmatrix}$$

- **Unknown vector:** $\beta \in \mathbb{R}^N$ i.i.d. with distribution p_β ,
- **Measurement matrix:** $A \in \mathbb{R}^{m \times N}$,
- **Measurement noise:** $w \in \mathbb{R}^m$ i.i.d. with variance σ^2 ,
- **Sampling ratio:** $\frac{m}{N} \in (0, \infty)$ constant, denoted δ .

High-Dimensional Regression

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$$\begin{bmatrix} y \end{bmatrix} = \begin{matrix} m \\ \left| \right. \end{matrix} \begin{matrix} \overbrace{\hspace{10em}}^N \\ \left[\hspace{10em} A \hspace{10em} \right] \end{matrix} \begin{bmatrix} \beta \end{bmatrix} + \begin{bmatrix} w \end{bmatrix}$$

Approximate Message Passing (AMP)

- AMP is a low complexity, scalable algorithm to solve the above regression task.
- AMP derived as approximation of loopy belief propagation for dense graphs
[Donoho-Maleki-Montanari '09], [Rangan '11], [Krzakala et al '12], [Schniter '11], ...

AMP Performance Guarantees

- AMP iteratively produces estimates of β denoted β^1, β^2, \dots
- Rigorous asymptotic analysis [Bayati-Montanari '11] when A is Gaussian:

$$\text{For each } t, \quad \lim_{m \rightarrow \infty} \frac{1}{m} \|\beta - \beta^t\|^2 = \sigma_t^2.$$

- σ_t^2 can be computed via a scalar iteration — *state evolution*.

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- σ_t^2 can be computed via a scalar iteration — *state evolution*.

State evolution still accurate for finite, large N

For each t ,

$$P\left(\left|\frac{1}{m} \|\beta - \beta^t\|^2 - \sigma_t^2\right| \geq \epsilon\right) \leq K_t e^{-\kappa_t N \epsilon^2}.$$

t -dependent constants K_t, κ_t .

AMP Algorithm

Set $\beta^0 = 0$. For $t \geq 0$:

$$z^t = y - A\beta^t + \frac{z^{t-1}}{m} \sum_{i=1}^N \eta'_t \left([A^T z^{t-1} + \beta^{t-1}]_i \right),$$

$$\beta^{t+1} = \eta_t(\beta^t + A^T z^t),$$

- z^t is the 'modified residual' after step t
- η_t denoises the *effective observation* to produce β^{t+1}

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The *correction* term in z^t ensures that for large enough N :

$$\beta^t + A^T z^t \approx \beta + \tau_t Z \quad \text{where } Z \text{ is } \mathcal{N}(0, 1)$$

⇒ The *effective observation* $\beta^t + A^T z^t$ is true signal observed in independent Gaussian noise.

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- z^t is the 'modified residual' after step t
- η_t denotes the *effective observation* to produce β^{t+1}

- If p_β is known, the Bayes-optimal choice for η_t which minimizes $\mathbb{E}[\|\beta - \beta^{t+1}\|^2]$ is

$$\eta_t(s) = \mathbb{E}[\beta \mid \beta + \tau_t Z = s]$$

- If p_β is unknown, partial knowledge about β can guide the choice of η_t .

The Modified Residual [Donoho-Maleki-Montanari '09]

Assume $A_{ij} \sim \mathcal{N}(0, 1/m)$ and $w_i \sim \mathcal{N}(0, \sigma^2)$.

Suppose instead,

$$z^t = y - A\beta^t = w - A(\beta - \beta^t)$$

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Then effective observation:

$$\beta^t + A^T z^t = \beta + A^T w + (I - A^T A)(\beta - \beta^t)$$

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Then effective observation:

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- The 'correction' term asymptotically cancels out dependence between $(I - A^T A)$ and $(\beta - \beta^t)$ so that

$$\beta^t + A^T z^t \approx \beta + \tau_t Z_t, \quad \text{where } \tau_t^2 = \sigma^2 + \frac{1}{m} \mathbb{E} \|\beta - \beta^t\|^2$$

State Evolution

Define τ_t^2 as the variance of the noise in the effective observation after step t .

$$\beta^t + A^T z^t \approx \beta + \tau_t Z, \quad Z \sim \mathcal{N}(0, \mathbb{I}).$$

SE Equations:

Set $\tau_0^2 = \sigma^2 + \mathbb{E}\|\beta\|^2/m$,

$$\tau_t^2 = \sigma^2 + \frac{\mathbb{E}\|\beta - \beta^t\|^2}{m}.$$

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SE Equations:

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$$\tau_t^2 = \sigma^2 + \frac{\mathbb{E}\|\beta - \beta^t\|^2}{m} = \sigma^2 + \frac{\mathbb{E}\|\beta - \eta_t(\beta + \tau_{t-1}Z)\|^2}{m}.$$

$Z \sim \mathcal{N}(0, 1)$ independent of $\beta \sim p_\beta$.

Assumptions

We make the following assumptions:

- **Measurement matrix:** i.i.d. $\sim \mathcal{N}(0, 1/m)$.
- **Signal:** i.i.d. $\sim p_\beta$, sub-Gaussian.
- **Measurement noise:** i.i.d. $\sim p_w$, sub-Gaussian, $\mathbb{E}[w_i^2] = \sigma^2$.
- **De-noising Functions η_t :** Lipschitz continuous with weak derivative η'_t which is differentiable except possibly at a finite number of points, with bounded derivative everywhere it exists.

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Pseudo-Lipschitz (PL) Loss Functions

A function $\phi : \mathbb{R}^n \rightarrow \mathbb{R}$ is PL if there exists a constant $L > 0$ such that for all $x, y \in \mathbb{R}^n$,

$$|\phi(x) - \phi(y)| \leq L(1 + \|x\| + \|y\|)\|x - y\|.$$

E.g. $\phi(x) = \|x\|_2^2$ (l_2 loss) or $\phi(x) = \|x\|_1$ (l_1 loss).

Main Result

Theorem

Under the assumptions of the previous slide, for any PL function $\phi : \mathbb{R}^2 \rightarrow \mathbb{R}$, $\epsilon \in (0, 1)$, and $t \geq 0$,

$$P\left(\left|\frac{1}{m} \sum_{i=1}^N \phi(\beta_i^{t+1}, \beta_i) - \mathbb{E}[\phi(\eta_t(\beta + \tau_t Z), \beta)]\right| \geq \epsilon\right) \leq K_t e^{-\kappa_t N \epsilon^2},$$

for $Z \sim \mathcal{N}(0, 1)$, $\beta \sim p_\beta$ independent with constants K_t, κ_t .

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Choosing PL loss function $\phi(a, b) = (a - b)^2$, the Theorem proves

$$P\left(\left|\frac{1}{N} \|\beta^{t+1} - \beta_0\|^2 - \delta(\tau_{t+1}^2 - \sigma^2)\right| \geq \epsilon\right) \leq K_t e^{-\kappa_t N \epsilon^2},$$

for τ_{t+1}^2 computed via state evolution.

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for $Z \sim \mathcal{N}(0, 1)$, $\beta \sim p_\beta$ independent with constants K_t, κ_t .

- This refines an asymptotic result proved by Bayati, Montanari [Trans. IT '11]
- The finite-sample result above implies the asymptotic result (via Borel-Cantelli).

Main Result

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for $Z \sim \mathcal{N}(0, 1)$, $\beta \sim p_\beta$ independent with constants K_t, κ_t .

Constants in the Bound:

- Constants $K_t = K_1(K_2)^t(t!)^{10}$ and $\kappa_t = \kappa_1\kappa_2^{-t}(t!)^{-22}$ where $K_1, K_2, \kappa_1, \kappa_2 > 0$ are universal constants.
- Indicates how large t can get for deviation prob. $\rightarrow 0$:

$$t = o\left(\frac{\log N}{\log \log N}\right)$$

Main Result: Proof Sketch

Show $\beta^t + A^T z^t \sim \beta + \tau_t Z$, with τ_t given by state evolution.

Steps

1. Characterize the conditional distribution of the effective observation and residual as sum of i.i.d. Gaussians plus deviation term.

Show:

$$\begin{aligned}(\beta^t + A^T z^t - \beta)|_{\{\text{past}, \beta, w\}} &\stackrel{d}{=} \tau_t Z_t + \Delta_t, \\(z^t - w)|_{\{\text{past}, \beta, w\}} &\stackrel{d}{=} \sqrt{\tau_t^2 - \sigma^2} \tilde{Z}_t + \tilde{\Delta}_t,\end{aligned}$$

where Δ_t is complicated – constructed of inner products of other output from AMP iterations up to time t .

2. Inductively obtain concentration results showing that the deviation terms are small with high probability.

Role of the Correction Term

For Δ_t to concentrate at 0, correction term takes form $\gamma_{t-1}z^{t-1}$.

Ideal Value

For all $j \leq t - 1$, want *simultaneously*:

$$\gamma_{t-1} := \left(\frac{\tau_j}{\tau_{t-1}} \right) \frac{1}{\delta} \mathbb{E} [Z_j \eta_{t-1} (\beta + \tau_{t-1} Z_{t-1})]$$

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$$= \dots$$

$$= \frac{1}{\delta} \mathbb{E} [\eta'_{t-1}(\beta + \tau_{t-1} Z_{t-1})]$$

Stein's Lemma

$$\mathbb{E} [Z_j Z_{t-1}] = \frac{\tau_{t-1}}{\tau_j}$$

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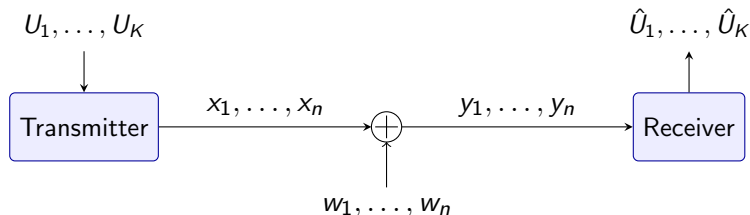
If we don't know p_β ...

Use inductive hypothesis $\beta^{t-1} + A^T z^{t-1} \approx \beta + \tau_{t-1}Z_{t-1}$ to estimate

$$\hat{\gamma}_{t-1} = \frac{1}{\delta N} \sum_{i=1}^N \eta'_{t-1}([\beta^{t-1} + A^T z^{t-1}]_i)$$

An Application: Sparse Regression Codes

The Additive White Gaussian Noise Channel



$$y_i = x_i + w_i, \quad i = 1, \dots, n$$

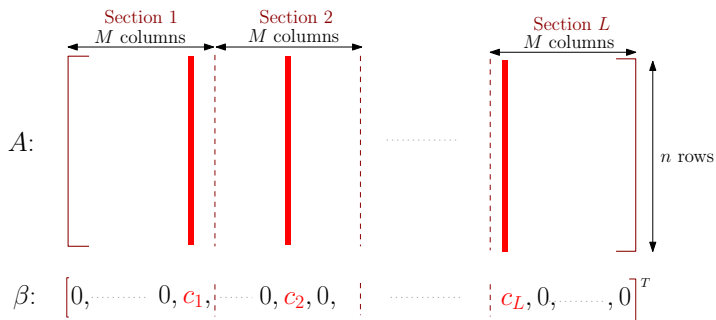
- Want to convey K bits in n channel uses
- Additive Gaussian noise: $w_i \text{ iid } \sim \mathcal{N}(0, \sigma^2)$
- Average power constraint: $\frac{1}{n} \sum_i x_i^2 \leq P$
- Rate: $R = K/n$ bits/transmission
- Capacity: $\mathcal{C} = \frac{1}{2} \log(1 + \text{snr})$

GOAL: Codes with fast encoding & decoding with
 $\text{Prob}(\hat{\underline{U}} \neq \underline{U}) \rightarrow 0$ at rates R approaching \mathcal{C}

Sparse Regression Codes (SPARCs)

- Introduced by Barron and Joseph & shown to be capacity achieving with feasible decoding ['10, '12]
- We study an *approximate message passing* decoder:
 - Low-complexity, provably capacity-achieving with near-exponential decay of error probability
 - Good performance at practical block lengths

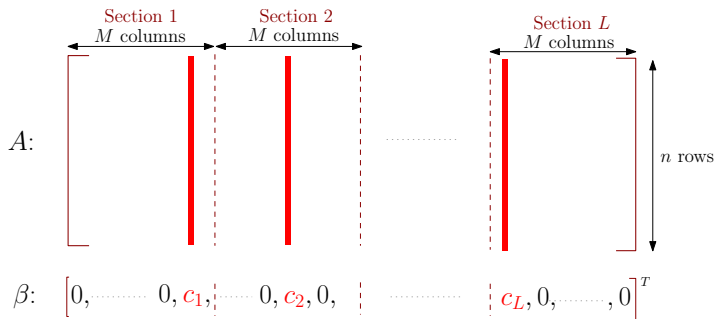
SPARC Codebook Construction



n rows, ML columns

- Codeword $A\beta$ is linear combinations of L columns of A .
- Message vector β has L entries with non-zero values c_1, c_2, \dots, c_L fixed.
- Message bits determine the location of the non-zeros.

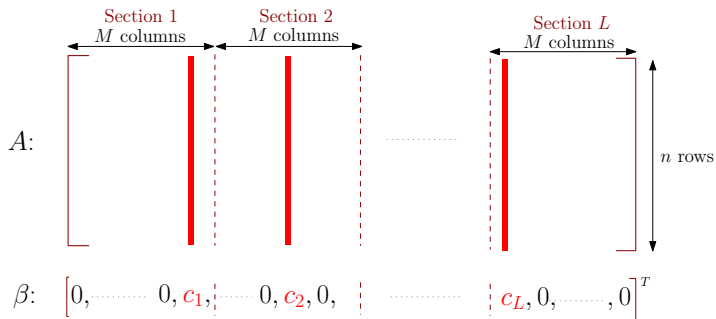
SPARC Codebook Construction



Choosing M and L :

- $M^L = 2^{nR}$ codewords for rate R , hence $L \log M = nR$
- Encoding: L chunks of input bits, each chunk with $\log M$ bits

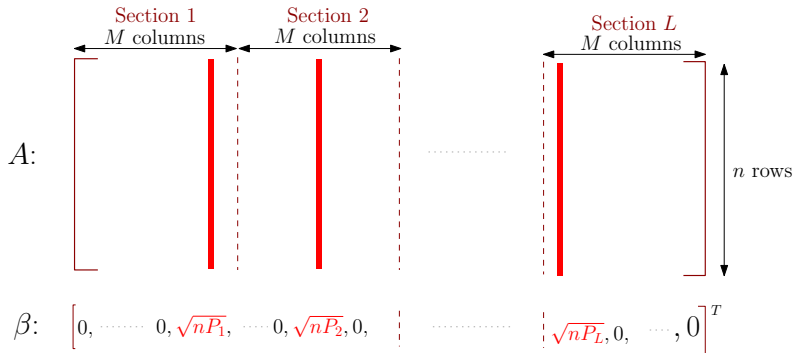
SPARC Codebook Construction



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- $M^L = 2^{nR}$ codewords for rate R , hence $L \log M = nR$
- Encoding: L chunks of input bits, each chunk with $\log M$ bits
- Choosing $M = L^a$, we get $L \sim n/\log n$
- Size of $A = n \times ML$, also **polynomial** in n

Power Allocation



$c_1 = \sqrt{nP_1}, \dots, c_L = \sqrt{nP_L}$ chosen such that $\sum_{\ell} P_{\ell} = P$

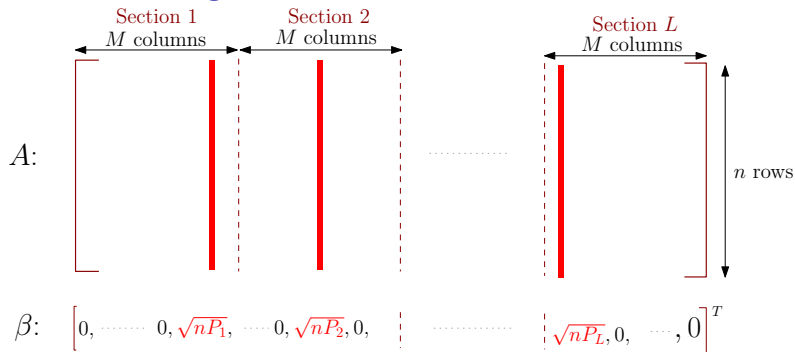
Examples

1. Flat: $P_{\ell} = \frac{P}{L}$

2. Exponentially Decaying: $P_{\ell} \propto e^{-\kappa\ell/L}$, constant $\kappa > 0$

For all power allocations, $P_{\ell} = \Theta(\frac{1}{L})$, $\sqrt{nP_{\ell}} = \Theta(\sqrt{\log M}) \quad \forall \ell$

Optimal Decoding



$$\text{Channel output: } y = A\beta + w$$

- $\hat{\beta}_{\text{opt}} = \arg \min_{\hat{\beta} \in \mathcal{B}} \|y - A\hat{\beta}\|^2$
- Probability of decoding error falls exponentially with n
Similar performance to Shannon-style random coding
[Joseph-Barron '12]
- Decoding complexity grows exponential with n

Feasible Decoders

- Adaptive successive decoders by [Joseph-Barron '12], [Barron-Cho '13] are asymptotically capacity-achieving
- **AMP decoding: [Rush, Greig, Venkataramanan '15], [Barbier, Krzakala '15]**
 - Asymptotically capacity-achieving [Rush, Greig, Venkataramanan '15]
 - Good performance at practical block lengths
 - Near-exponential decay of the error probability [Rush, Venkataramanan '17]
 - Finite sample analysis gives guidance on how to choose code parameters at finite block lengths..

Performance of AMP Decoder

Some details I'm leaving out...

- Derivation of AMP algorithm for SPARCs problem.
- Prove state evolution accurately predicts performance.

State evolution tells us:

- Run AMP decoder for T^* iterations, where

$$T^* := 1 + \left\lceil \left(\frac{1}{2C} \log \left(\frac{C}{R(1 + \alpha/2)} \right) \right)^{-1} \right\rceil,$$

$\alpha \in [0, 1/2]$ constant.

- After $T^* - 1$ iterations, we are guaranteed that

$$\sigma^2 \leq \tau_{T^*}^2 \leq \sigma^2 + \kappa M^{-c\alpha^2}.$$

- At final iteration T^* , generate

$$\beta^{T^*} = \eta_{T^*} \left(\underbrace{A^T z^{T^*-1} + \beta^{T^*-1}}_{\approx \beta + \sigma Z} \right).$$

Performance of AMP Decoder

The *section error rate* of a decoder for SPARC \mathcal{S} is

$$\mathcal{E}_{\text{sec}}(\mathcal{S}) := \frac{1}{L} \sum_{\ell=1}^L \mathbf{1}\{\hat{\beta}_{\ell} \neq \beta_{\ell}\}.$$

Theorem: [Rush, Greig, Venkataramanan '15]

Fix any rate $R < \mathcal{C}$ and $b > 0$. Consider a SPARC \mathcal{S}_n with rate R , block length n , design matrix parameters L and $M = L^b$ determined according to

$$L \log M = nR,$$

and exponentially decaying power allocation, $P_{\ell} \propto 2^{-2\mathcal{C}\ell/L}$.

Then the section error rate of the AMP decoder converges to zero almost surely, i.e., for any $\epsilon > 0$,

$$\lim_{n_0 \rightarrow \infty} P(\mathcal{E}_{\text{sec}}(\mathcal{S}_n) < \epsilon, \forall n \geq n_0) = 1.$$

Performance of AMP Decoder

Theorem: [Rush, Venkataramanan '17]

Fix any rate $R < \mathcal{C}$. Consider a SPARC \mathcal{S}_n with rate R , block length n , design matrix parameters L, M determined according to

$$L \log M = nR,$$

and an exponentially decaying power allocation with $P_\ell \propto 2^{-2\mathcal{C}\ell/L}$. Let $\epsilon > \kappa M^{-c\alpha^2}$. Then, the section error rate of the AMP decoder satisfies

$$\mathbb{P}(\mathcal{E}_{\text{sec}}(\mathcal{S}_n) > \epsilon) \leq K_{T^*} \exp \left\{ \frac{-\kappa_{T^*} L \epsilon^2}{(\log M)^{2T^*-1}} \right\},$$

- Constants K_{T^*}, κ_{T^*} depend only on T^* , no further dependence on n, M, L, ϵ .
- The dependence on the rate R is only via T^* :
Recall $T^* \approx 1 + 2\mathcal{C} / \log(\frac{\mathcal{C}}{R})$.

Error Exponent

Consider the choice $M = L^a$ for constant $a > 0 \Rightarrow L \sim \frac{n}{\log n}$

For this choice:

$$\mathbb{P}(\mathcal{E}_{\text{sec}}(\mathcal{S}_n) > \epsilon) \leq K_{T^*} \exp \left\{ \frac{-n\kappa_{T^*}\epsilon^2}{(\log n)^{2T^*}} \right\}$$

- Can also consider other choices such as
 $L \sim \frac{n}{\log \log n}, \quad M \sim \log n$
- Leads to different trade-offs with respect to error exponent, complexity, and gap-to-capacity

Performance of AMP Decoder

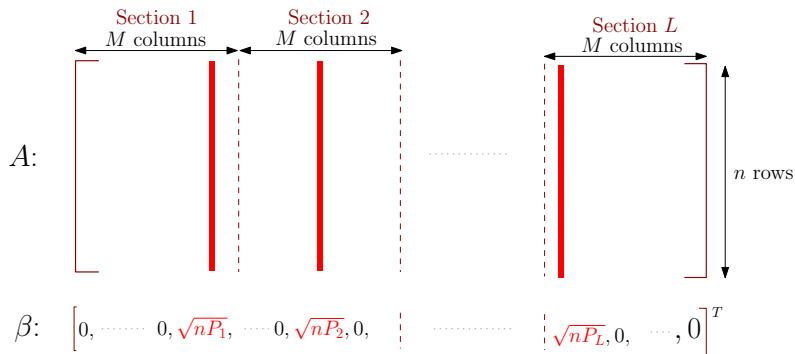
Finite sample analysis allows us to analyze the gap to capacity Δ_R :

- For fixed section error rate ϵ , how fast can R approach \mathcal{C} with growing n ?
- For the choice $M = L^a$,

$$\Delta_R \text{ is of order } \sqrt{\frac{\log \log n}{\log n}}.$$

- For AWGN channels, no known coding scheme that provably achieves a polynomial gap to capacity with efficient decoding.

Choosing L, M



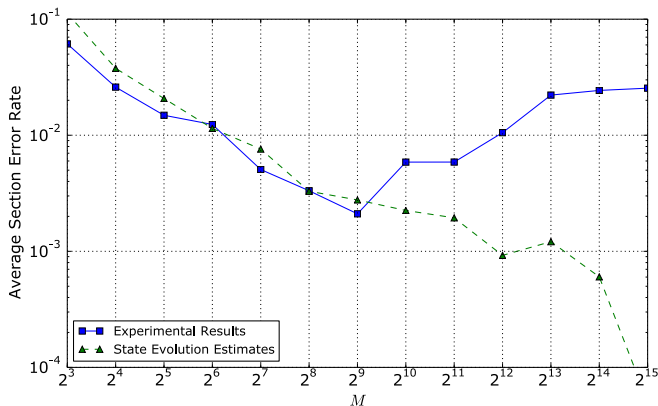
$$L \log M = nR$$

Larger $L, M \Rightarrow$ better performance?

Let's try increasing M with L fixed

Choosing L, M

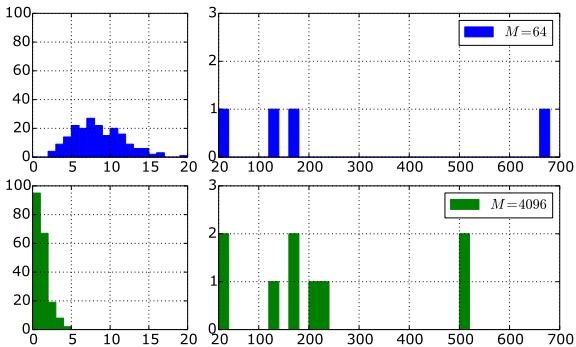
- State evolution prediction: $0 \leq \frac{1}{n} \mathbb{E} \|\beta - \beta^T\|^2 \leq \kappa M^{-c\alpha^2}$
- So we might expect performance to improve with $\uparrow M$



AMP performance with increasing M , for $L = 1024$, $R = 1.5$, $\frac{E_b}{N_0} = 5.7\text{dB}$
[Greig and Venkataramanan '17]

$$\mathbb{P}(\mathcal{E}_{\text{sec}}(\mathcal{S}_n) > \epsilon) \leq K_{T^*} \exp \left\{ \frac{-\kappa_{T^*} L \epsilon^2}{(\log M)^{2T^*-1}} \right\},$$

- Larger $M \Rightarrow$ worse concentration (with L, R, snr fixed)
- T^* is of the order of tens, so this effect is significant!



$M = 64$ vs. $M = 4096$: Histogram of section errors

Increasing L improves concentration
 Increasing M useful up to a point

Summary

Finite Sample AMP

- SE accurate in predicting AMP performance for large, finite N .
- Prob. of deviation from SE predictions falls exponentially like $N\epsilon^2$.
- Theoretical support for empirical findings of such.

Future directions:

- Extensions to LASSO risk, high-dimensional M-estimation, low-rank matrix estimation, AMP with spatially coupled matrices, and GAMP?
- Theoretical results for general A matrices (iid uniform Bernoulli, partial DFT, . . .)

Summary

AMP for SPARCs

- AMP for low-complexity SPARC decoding.
- Any rate $R < \mathcal{C}$, probability of section error rate $> \epsilon$ decays exponentially as

$$L\epsilon^2/(\log M)^{T^*} \approx n\epsilon^2/(\log n)^{T^*}.$$

Future directions:

- Theoretical Guarantees for the Hadamard-based SPARC.
- Theoretical analysis of spatially-coupled design matrices to improve performance at finite block-lengths.
- Will combining power allocation & spatial coupling give better empirical performance at reasonable block lengths near \mathcal{C} ?
- The BIG question: Can we design feasible decoders with $O(1/n^\alpha)$ gap to capacity for some $\alpha \in (0, 1/2)$?